

**IFPAC 2003, Scottsdale, AZ, January 21-24, 2003**  
**Session: Process Modeling**

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**Identification of Chemical Process  
Systems Using Constrained  
Minimum Crest Factor Multisine Inputs**

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**Presentation Outline**

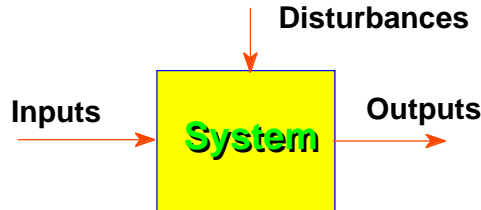
- What constitutes "plant-friendly" system identification?
- Minimum Crest Factor Multisine Signal Design
  - Schroeder (1970) phases
  - Guillaume *et al.* (1991) algorithm
  - Constrained problem formulation and solution
- Problem 1: Single-input, single-output (SISO) problem meaningful to control-relevant identification.
- Problem 2: SISO problem meaningful to nonlinear system identification
- Problem 3: Multi-input, multi-output (MIMO) problem, using a highly interactive 2x2 plant example.
- Summary and Conclusions



## System Identification

*"Identification is the determination, on the basis of input and output, of a system within a specified class of systems, to which the system under test is equivalent."*

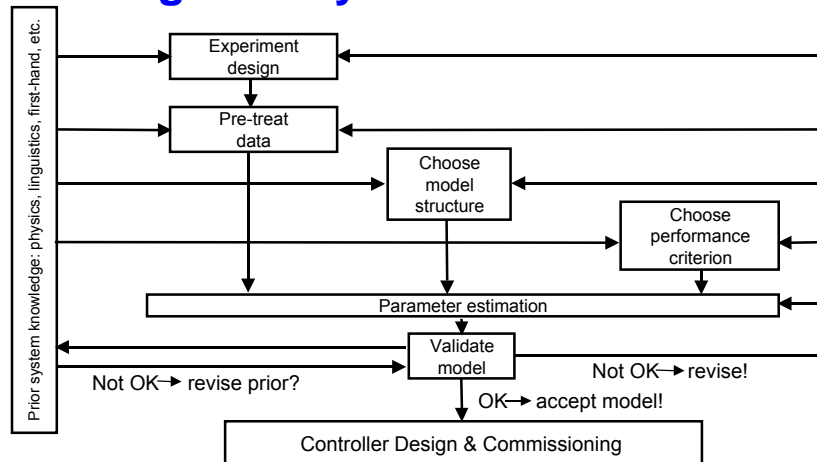
- L. Zadeh, (1962)



System identification focuses on the modeling of dynamical systems from experimental data



## Stages of System Identification



• from P. Lindskog, ISY, Linköping University, Sweden



## "Plant Friendly" Input Signal Design

A plant friendly input signal should:

- be as short as possible
- not take actuators to limits, or exceed move size restrictions
- cause minimum disruption to the controlled variables (i.e., low variance, small deviations from setpoint)

*Theoretical requirements, however, strongly conflict with "plant-friendly" operation!*



## Multisine Input Signals

A multisine input is a deterministic, periodic signal composed of a harmonically related sum of sinusoids,

$$u_s(k) = \lambda \sum_{i=1}^{n_s} \sqrt{2\alpha_i} \cos(\omega_i kT + \phi_i)$$

$T \equiv$  sampling time

$N_s \equiv$  sequence length

$n_s \equiv$  no. of sinusoids,  $n_s \leq N_s/2$

$\phi_i \equiv$  phase angle for harmonic  $i$

$\alpha_i \equiv$  relative power  $\left( \sum_{i=1}^{n_s} \alpha_i = 1 \right)$

$\omega_i = 2\pi i / N_s T$

$\lambda \equiv$  scaling factor to insure signal meets  $\pm u_{sat}$

The choice of phases  $\phi_i$ ,  $i = 1, \dots, n_s$  strongly influences the time-domain realization of the signal.



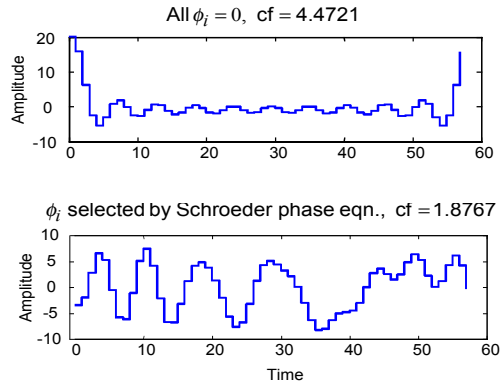
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The Crest Factor ( $CF$ ) of a signal  $u$  is the ratio of the  $\ell_\infty$  (or Chebyshev) norm and the  $\ell_2$ -norm

$$CF(u) = \frac{\ell_\infty(u)}{\ell_2(u)}$$

A low crest factor indicates that most of the elements in the input sequence are distributed near the minimum and maximum values of the sequence.



## Schroeder Phases

The work of Schroeder (1970) (Schroeder, M.R. (*IEEE Trans. Information Theory*, **16**, 85, Jan. 1970) presents a closed-form formula to select the phases in a multisine signal

$$\phi_i = 2\pi \sum_{j=1}^i j\alpha_j$$

The formula gives a good result when the user-defined spectrum is flat and wideband, but under other conditions (bandlimited, in the presence of harmonic suppression, etc.) the results can be very undesirable.



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## Research Goal

- Develop a design methodology for multisine inputs that incorporates *both* time-domain "plant-friendliness" constraints and frequency-domain information-theoretic requirements.
- Information-theoretic requirements include such considerations as:
  - persistence of excitation
  - control-relevance
  - harmonic suppression (to address nonlinearity)
  - orthogonality (for multi-input implementation)



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## Problems with Crest Factor Minimization

- Nonlinear optimization problem
- Nonsmooth (as a consequence of the  $\mathcal{L}$ -infinity norm)
- Nonconvex objective function

*Problems are worsened as a result of increasing problem dimension*



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### **Guillame *et al.* (1991) Algorithm**

The method of Guillaume *et al.* (1991)\* approximates the minimization of the  $\ell_\infty$  norm by sequentially minimizing the  $\ell_p$  norm for  $p = 4, 8, 16, \dots$ . It is based on Pólya's algorithm which states that

$$\lim_{p \rightarrow \infty} \mathbf{p}_p = \mathbf{p}_\infty$$

where  $\mathbf{p} = [\phi_1 \ \phi_2 \ \phi_3 \ \dots \ \phi_{n_s}]$  is the real-valued phase vector for

$$u(k) = \lambda \sum_{i=1}^{n_s} \sqrt{2\alpha_i} \cos(\omega_i k T + \phi_i)$$

and  $\mathbf{p}_\infty$  is the minimax solution. Since the  $\ell_2$ -norm remains invariant with respect to the phases  $\phi_i$ , this method effectively approximates the minimization of the crest factor (*CF*).

\*Guillaume, P., J. Schoukens, R. Pintelon and I. Kollár (1991). Crest-factor minimization using nonlinear chebyshev approximation methods. *IEEE Trans. on Inst. and Meas.*, **40**(6), 982-989.



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### **Constrained Solution Approach**

Our solution approach is similar to Guillaume *et al.* (1991) and is based on Pólya's algorithm.

- The problem is formulated in the modeling language AMPL, which provides exact, automatic differentiation up to second derivatives.
- A more gradual increase of  $p$  is performed in comparison to Guillaume *et al.* (1991).

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## Constrained Solution Approach (Continued)

- The trust region, interior point method developed by Nocedal and co-workers (Byrd, R., M.E. Hribar, and J. Nocedal. “An interior point method for large scale nonlinear programming.” *SIAM J. Optim.* 9, pgs 877–900, 1999) was applied in our initial research.
  - In this talk, improved results based on the Filter-SQP algorithm developed by Fletcher and co-workers (Fletcher, Leyffer, and Toint, “On the Global Convergence of a Filter-SQP Algorithm,” Numerical Analysis Report NA/197, University of Dundee, October 2000) are presented.
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## Problem 1

Given the multisine signal structure

$$u_s(k) = \lambda \sum_{i=1}^{n_s} \sqrt{2\alpha_i} \cos(\omega_i k T + \phi_i)$$

and a known power spectral density (defined by the Fourier coefficients  $\lambda\sqrt{2\alpha_i}$  for  $n_s$  spectral lines), solve the optimization problem

$$\min_{[\phi_1 \ \phi_2 \ \dots \ \phi_{n_s}]} \text{CF}(u_s)$$

subject to maximum move size constraints on the input,

$$|\Delta u_s(k)| \leq \Delta u^{max} \quad \forall k$$

and possibly high/low limits on  $u(k)$ ,

$$u^{min} \leq u_s(k) \leq u^{max} \quad \forall k$$



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Consider a first-order with deadtime plant model

$$\frac{y(s)}{u(s)} = p(s) = \frac{Ke^{-\theta s}}{\tau s + 1} = \frac{e^{-3s}}{3s + 1}, \quad T = 1 \text{ min}$$

The power spectrum for the input is chosen based on the control-relevant prefilter result presented in Rivera *et al.* 1992\*:

$$L(z) = \tilde{p}_e(z)\tilde{p}^{-1}(z)\tilde{\epsilon}(z)\tilde{\eta}(z)(r(z) - d(z))$$

$\tilde{p}(z) \equiv$  plant model

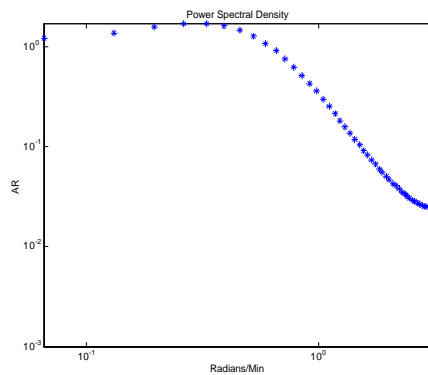
$\tilde{p}_e(z) \equiv$  noise model

$\tilde{\eta}(z) = \tilde{p}c(1 + \tilde{p}c)^{-1}$  complementary sensitivity

$\tilde{\epsilon}(z) = (1 + \tilde{p}c)^{-1}$  sensitivity

$r(z) - d(z) \equiv$  setpoint/disturbance direction

\*Rivera, D.E., J.F. Pollard, and C.E. García, (1992) "Control-Relevant Prefiltering: A Systematic Design Approach and Case Study," *IEEE Trans. Autom. Cntrl.*, **37**, 964-974.



$$u(z) = (z - \alpha)(1 - z^{-nk}f(z))z^{-1}f(z)(r(z) - d(z)), \quad f(z) = \frac{(1 - \delta)^2 z^2}{(z - \delta)^2}$$

$\delta = e^{-1.555T/\tau_{cl}}$ ;  $\tau_{cl} = 2.25$  min is the desired closed-loop time constant.

$\alpha = e^{-T/\tau_{dom}}$ ;  $\tau_{dom} = 4.5$  min is a dominant time constant estimate

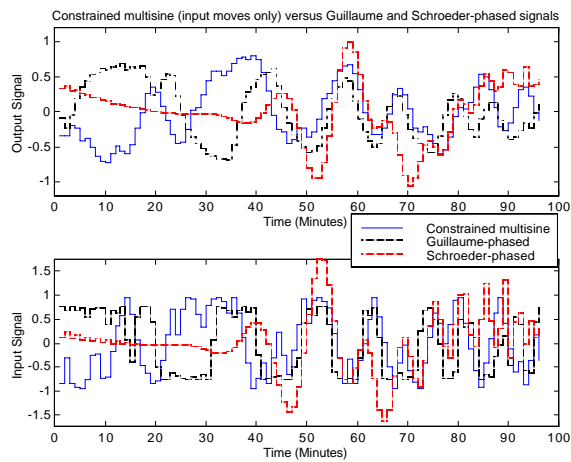


**Problem 1 Example Results**

Signal	CF( <i>u</i> )	max( $\Delta u$ )	min( <i>u</i> )	max( <i>u</i> )	CF( <i>y</i> )	max( $\Delta y$ )	min( <i>y</i> )	max( <i>y</i> )
Schroeder-phased	2.7966	1.6214	-1.6384	1.75	2.5892	0.4846	-1.0524	0.9984
Guillaume-phased	1.2173	1.1013	-0.7618	-0.7612	1.7020	0.3506	-0.6918	0.6835
min CF( <i>u</i> ) Constrained ( $\Delta u^{max} = 0.525$ )	1.5051	0.525	-0.9417	0.9413	1.9658	0.3487	-0.7280	0.7990
min CF( <i>y</i> ) Unconstrained	2.4036	1.3119	-1.3778	1.5052	1.2032	0.4405	-0.4891	0.4861
min CF( <i>y</i> ) ( $\Delta u^{max} = 0.525$ ; $\Delta y^{max} = 0.4$ )	2.039	0.525	-1.2758	1.2632	1.3860	0.3300	-0.5601	0.5634

**- Maximum move size reduced by 50% with only a relatively low (~24% increase) in crest factor and overall input signal span**

**Constrained min CF vs Guillaume and Schroeder-phased signals**



Constrained multisine input signal (with  $\Delta u^{max} = 0.525$ ) compared to Guillaume-phased and Schroeder-phased signals.

### Problem 1(b)

Minimize the crest factor of the plant *output*

$$\min_{[\phi_1 \ \phi_2 \ \dots \ \phi_{n_s}]} \text{CF}(y)$$

subject to maximum move size constraints on the input

$$|\Delta u_s(k)| \leq \Delta u^{max} \quad \forall k$$

output variability constraints

$$|\Delta y(k)| \leq \Delta y^{max} \quad \forall k$$

and high/low limits on  $u_s(k)$  and  $y(k)$

$$u^{min} \leq u_s(k) \leq u^{max} \quad \forall k$$

$$y^{min} \leq y(k) \leq y^{max} \quad \forall k$$

This scenario assumes that some *a priori* knowledge of the model is available to the optimizer to generate predicted outputs.



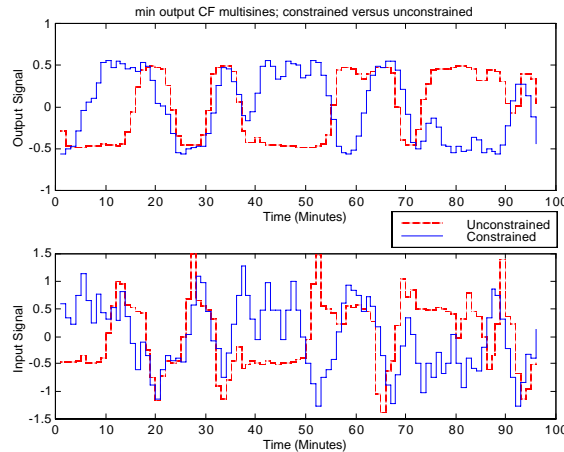
### Problem 1 Example Results

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Schroeder-phased	2.7966	1.6214	-1.6384	1.75	2.5892	0.4846	-1.0524	0.9984
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**Constrained problem solution sharply reduces the input move size and output changes. There is an unavoidable increase in the output CF, but the input CF and input/output spans are brought within acceptable levels.**

**Minimum Output CF signals (constrained vs unconstrained)**



Constrained multisine input signal (with  $\Delta u^{max} = 0.525$ ,  $\Delta y^{max} = 0.4$ ) compared to unconstrained multisine result (output crest factor minimized).

**Problem 2**

Given the multisine signal structure below and a known power spectrum (defined by the Fourier coefficients  $\lambda\sqrt{2\alpha_i}$  for  $n_s$  spectral lines)

$$u_s(k) = \lambda \sum_{i=1}^{n_s} \sqrt{2\alpha_i} \cos(\omega_i kT + \phi_i) + \sum_{i=n_s+1}^{n_a+n_s} \hat{a}_i \cos(\omega_i kT + \phi_i^a)$$

where  $n_a + n_s \leq N_s/2$  and  $\hat{a}_i$  and  $\phi_i^a$  represent the Fourier coefficients and phases, respectively, for  $n_a$  spectral lines beyond those defined in the user-specified spectrum, solve the optimization problem

$$\min_{[\phi_1 \ \phi_2 \ \dots \ \phi_{n_s}], [\phi_{n_s+1}^a \ \phi_{n_s+2}^a \ \dots \ \phi_{n_s+n_a}^a], [\hat{a}_{n_s+1} \ \hat{a}_{n_s+2} \ \dots \ \hat{a}_{n_s+n_a}]} CF(u_s)$$

subject to maximum move size constraints on the input,

$$|\Delta u_s(k)| \leq \Delta u^{max} \quad \forall k$$

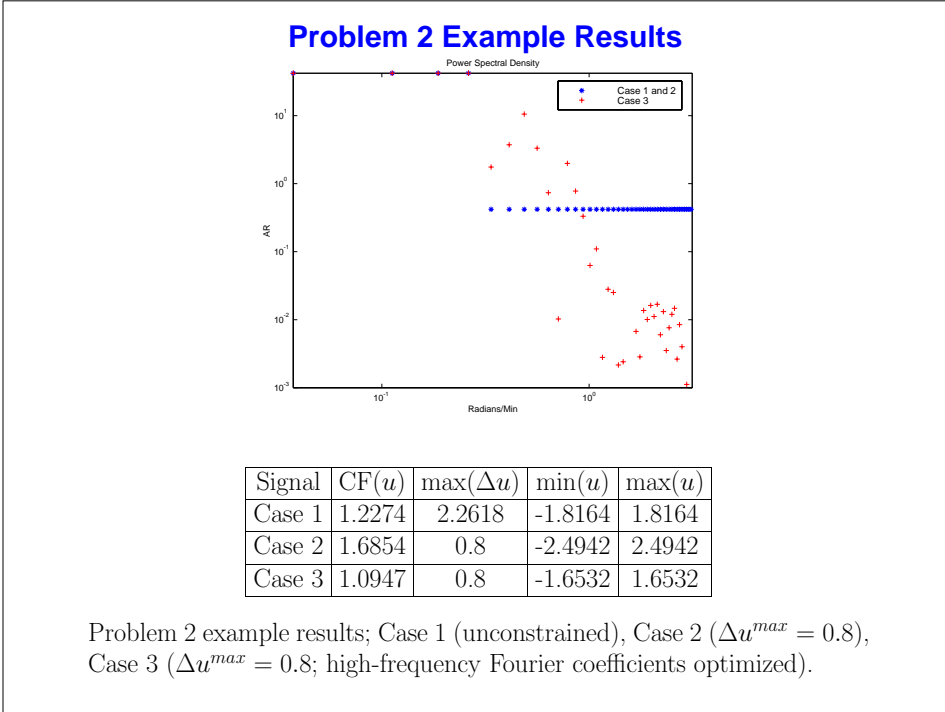
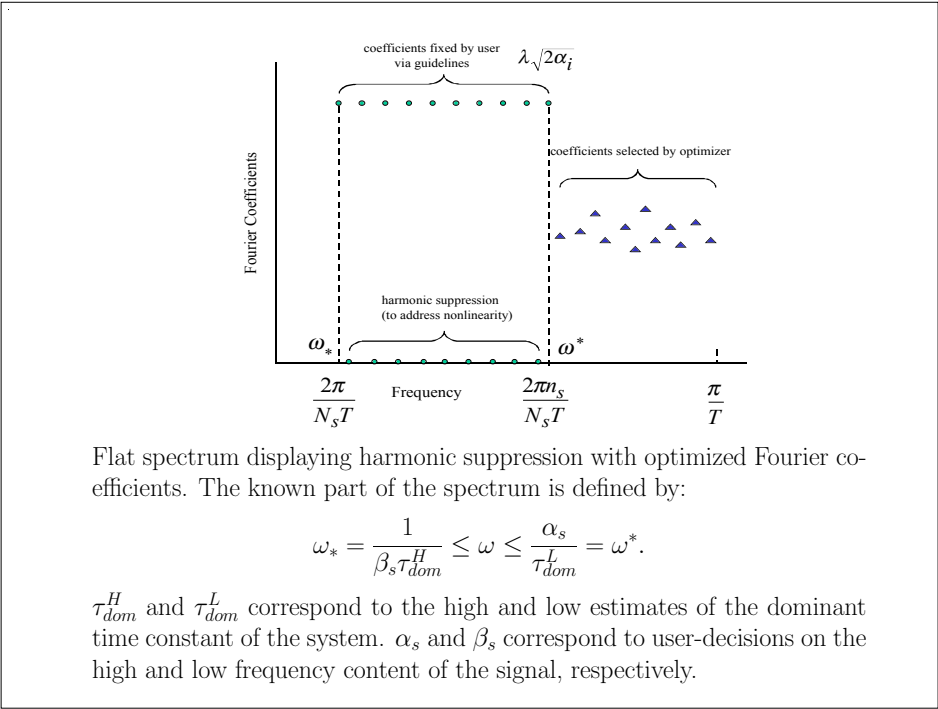
and possibly high/low limits on  $u(k)$ ,

$$u^{min} \leq u_s(k) \leq u^{max} \quad \forall k$$

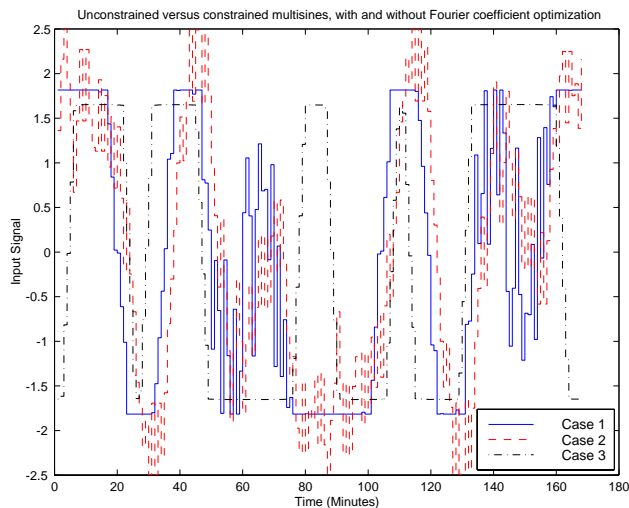


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## Problem 2 Example Results (Cont.)



Problem 2 example results; Case 1 (unconstrained), Case 2 ( $\Delta u^{max} = 0.8$ ), Case 3 ( $\Delta u^{max} = 0.8$ ; high-frequency Fourier coefficients optimized).

## Problem 3: Highly Interactive Multivariable System

$$P(s) = \frac{1}{75s + 1} \begin{bmatrix} 87.8 & -86.4 \\ 108.2 & -109.6 \end{bmatrix}$$

- Highly interactive (i.e., ill-conditioned) plants represent challenging problems for identification and control
- "Model 0" from Morari and Zafiriou, Robust Process Control, Prentice-Hall, (1987) has been studied.
- Simplest *meaningful* highly interactive problem we could find...

## Highly Interactive Systems

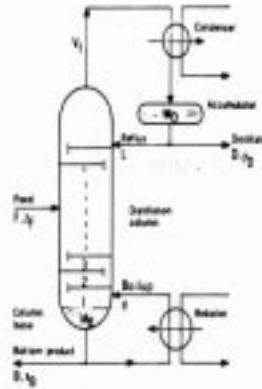


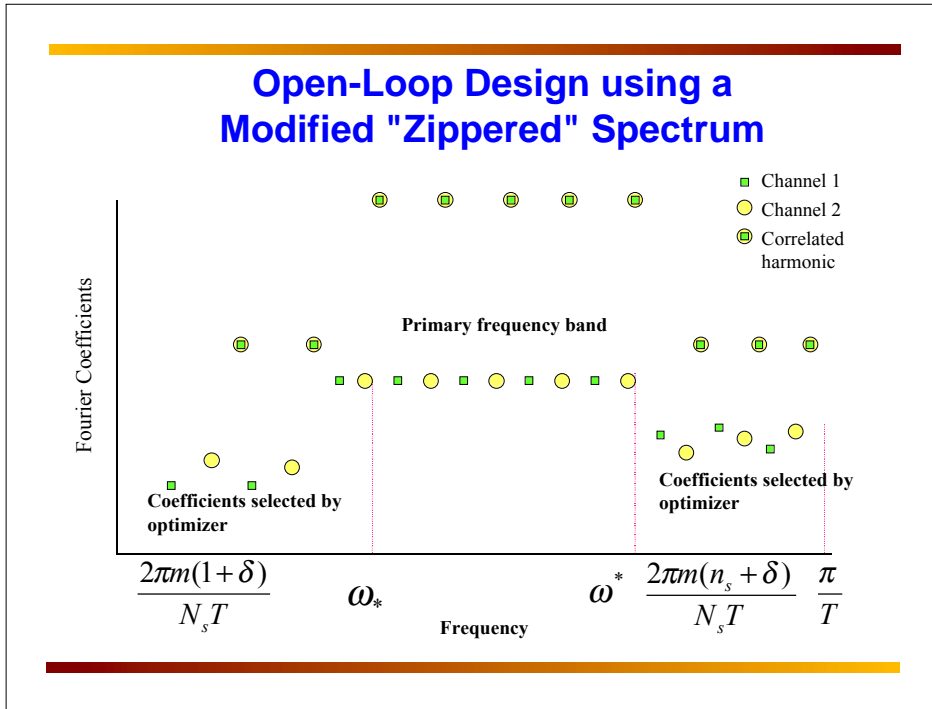
Fig. 2. Two-product distillation column.

*Highly interactive plants (such as high purity two-point binary distillation) present some inherent challenges to control design*

## Constrained Multisine Approaches for Highly Interactive Systems

- Open-loop design using a modified "zippered" spectrum. Overlapping/correlated sinusoids are introduced in the signal spectrum; these emphasize low-gain directionality in the data.
- Closed-loop design: Multisine signals are introduced as reference setpoints to the process under some form of closed-loop control.



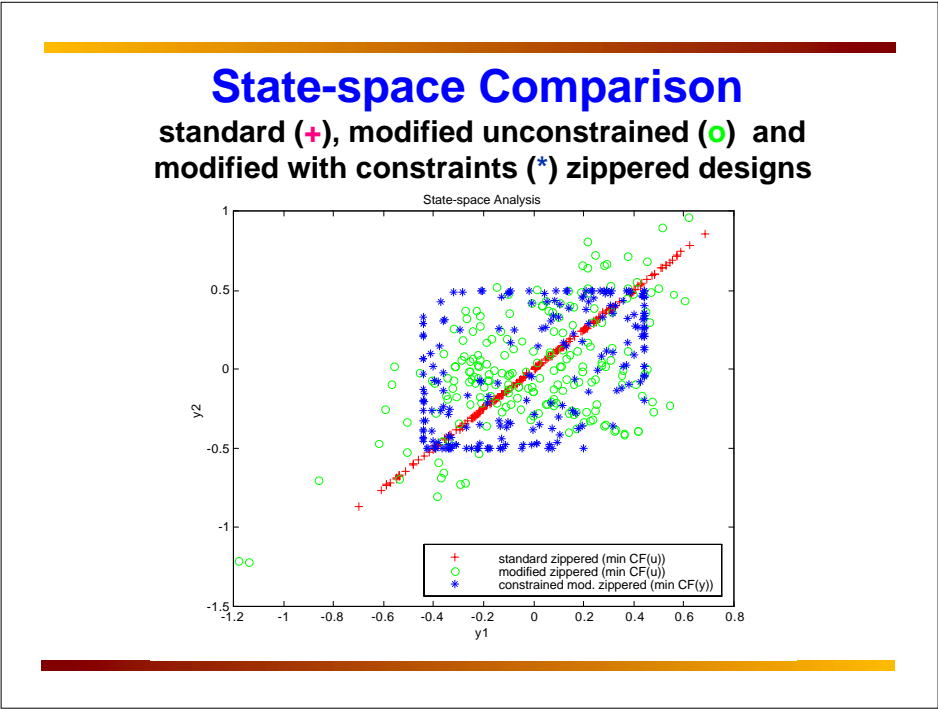


### Open-Loop Signals Summary

Type	Signal ( $x$ )	CF( $x$ )	PIPS(%)	max $\Delta x$	max $x$	min $x$
min CF ( $u$ ) design; standard zippered spectrum	$u_1$	1.132425	88.306039	0.014106	0.007079	-0.007079
	$u_2$	1.136376	87.999046	0.014201	0.007100	-0.007100
	$y_1$	2.383208	42.520481	0.318468	0.681899	-0.700354
	$y_2$	2.372958	42.530317	0.397652	0.855283	-0.871213
min CF ( $u$ ) design; modified zippered spectrum	$u_1$	1.130450	88.460763	0.745004	0.372522	-0.372526
	$u_2$	1.130632	88.446404	0.745163	0.372585	-0.372583
	$y_1$	3.924442	33.358140	0.763222	0.619802	-1.174440
	$y_2$	3.459978	32.254284	0.863657	0.966114	-1.219642
min CF ( $y$ ) design; modified zippered, $(\Delta u, \Delta y, y)^{max} \leq 0.5$ , $y^{min} > -0.5$ .	$u_1$	1.423676	70.240749	0.499997	0.464209	-0.464209
	$u_2$	1.423675	70.240777	0.499995	0.464217	-0.464217
	$y_1$	1.423637	70.242767	0.407867	0.441263	-0.441262
	$y_2$	1.363417	73.346052	0.499934	0.499980	-0.499967

Performance Index for Perturbation Signals (PIPS):

$$PIPS(\%) = 200 \frac{\sqrt{x_{rms}^2 - x_{mean}^2}}{x_{max} - x_{min}}$$



### Summary and Conclusions

- A constrained problem formulation allowing the user to specify desirable frequency-domain and time-domain specifications in low crest factor multisines has been presented.
- Its usefulness has been demonstrated in a series of problems ranging from signals for control-relevant identification to input signal design for highly interactive multivariable systems.
- ***The effective use of a priori knowledge is critical in the solution of this (or any other) control-oriented, plant-friendly identification problem.***





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### **References**

- Rivera, D.E., M.W. Braun, and H.D. Mittelmann, "Constrained multisine inputs for plant-friendly identification of chemical processes," 15th IFAC World Congress, Barcelona, Spain, July 21-26, 2002.
- Rivera, D.E., H. Lee, M.W. Braun, and H.D. Mittelmann, "Identification of multivariable process systems using constrained minimum crest factor multisine inputs," Paper 255f, 2002 AIChE Annual Meeting, Indianapolis, IN, Nov. 3-8, 2002.

Preprints of both papers can be downloaded from the CSEL website, <http://www.eas.asu.edu/~csel/pubs.htm>

